

BP3105

High Precision PSR CC LED Driver Controller

Description

BP3105 is a high precision primary side regulation (PSR) controller designed for high brightness LED driver. It's extremely suitable to be used in flyback converter operating in worldwide voltage range.

BP3105 works in discontinuous inductive current mode. The patented control method dramatically saves the system cost and size by eliminating all secondary feedback circuitry and loop control compensation circuitry. This control method also achieves excellent line and load regulation.

BP3105 integrates high precision current sense circuitry, which achieves $\pm 3\%$ LED current accuracy.

BP3105 offers rich protection features including LED short/open circuit protection, over-temperature protection, under-voltage lockout, VCC over voltage protection.

BP3105 is assembled in tiny SOT23-5 package.

Features

- ◆ PSR constant current (CC) control, no secondary feedback circuit required
- ◆ $\pm 3\%$ LED current accuracy
- ◆ Tiny SOT23-5 packagez
- ◆ Ultra low operating current to improve efficiency
- ◆ Low power consumption on FB resistor network
- ◆ World wide voltage range
- ◆ LED short/open circuit protection
- ◆ VCC over voltage protection
- ◆ Current sense resistor open circuit protection
- ◆ Over temperature protection
- ◆ No external loop compensation required

Applications

- ◆ GU10/E27 LED bulb, spot light
- ◆ LED PAR30/PAR38, down light
- ◆ LED tube light
- ◆ Other LED lighting

Typical Application

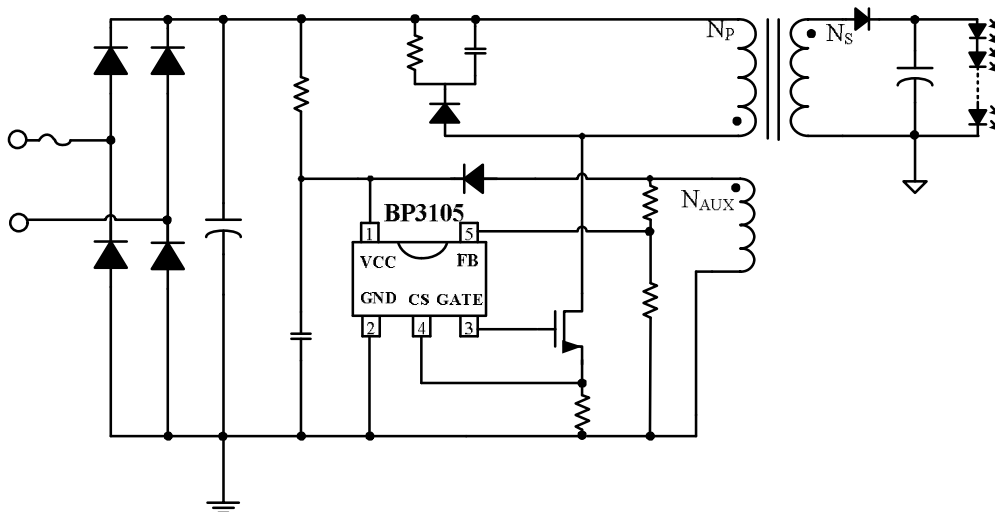
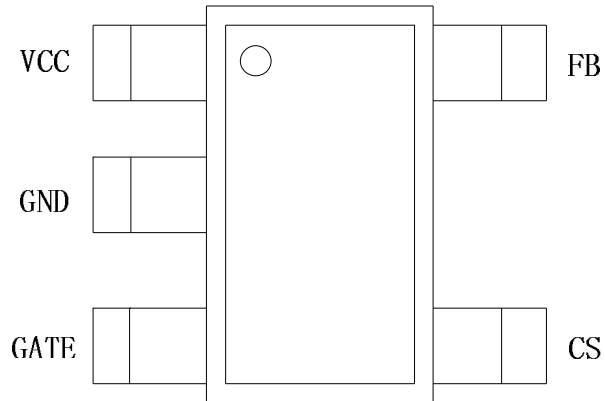


Figure 1 typical application circuit

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Pin Configuration



Pin Description

Pin No.	Pin Name	Description
1	VCC	Power supply
2	GND	Signal and Power ground
3	GATE	The output gate driver for an external N-channel power MOSFET.
4	CS	The current sense pin used to sense the MOSFET current by means of an external sense resistor. The resistor is connected between CS and GND
5	FB	The voltage feedback from auxiliary winding.

Ordering Information

Order Number	Package	Operating Temperature	Packing Method	Marking
BP3105ESOE	SOT23-5	-40 °C to 85 °C	Tape and Reel 3,000 units	3105X (note1)

Note1: X is stand for A,B,C,D

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Absolute Maximum Ratings (note2)

Symbol	Parameters	Value	Unit
V _{CC}	Power supply	-0.3~25	V
FB	The voltage feedback from auxiliary winding.	-0.3~6	V
CS	Current sense	-0.3~6	V
Gate	Gate driver of external NMOS	-0.3~25	V
P _{DMAX}	Power dissipation (note3)	0.3	W
P _{TR}	Thermal resistance, SOT-23-5 (θ_{JA})	300	°C/W
T _J	Operating temperature	-40 to 150	°C
T _{STG}	Storage temperature	-55 to 150	°C
	ESD (note4)	2	kV

Note 2: Stresses beyond those listed under “absolute maximum ratings” may cause permanent damage to the device. Under “recommended operating conditions” the device operation is assured, but some particular parameter may not be achieved. The electrical characteristics table defines the operation range of the device, the electrical characteristics is assured on DC and AC voltage by test program. For the parameters without minimum and maximum value in the EC table, the typical value defines the operation range, the accuracy is not guaranteed by spec.

Note 3: The maximum power dissipation decrease if temperature rise, it is decided by T_{JMAX} , θ_{JA} and environment temperature (T_A). The maximum power dissipation is the lower one between $P_{DMAX} = (T_{JMAX} - T_A) / \theta_{JA}$ and the number listed in the maximum table.

Note 4: Human Body mode, 100pF capacitor discharge on 1.5k Ω resistor

Recommended Operation Condition

Symbol	Parameters	Value	Unit
V _{CC}	Power supply voltage	7.5 ~14.5	V

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Electrical Characteristics ($V_{CC}=12V$, $T_A=25^\circ C$ unless otherwise stated)(note 5,6)

Symbol	Parameter	Conditions	Min.	Typ.	Max.	Unit
Supply voltage section						
V_{CC}	Input voltage		6.5	12	16	V
V_{CC-TH}	Start-up voltage	V_{CC} rising	13	14	15	V
V_{UVLO}	Under-voltage lockout threshold			6.5		V
V_{OVP}	Over voltage protection threshold			16		V
$V_{CC-CLAMP}$	V_{CC} clamp voltage	$I_{VCC}=10mA$		19		V
Current sense section						
V_{CS-TH}	Current sense threshold		495	500	505	mV
T_{LEB}	LEB time			500		ns
T_{DELAY}	Switch off delay			200		ns
Operation current section						
I_{ST}	Start up current	$V_{CC}=V_{CC-TH} - 1V$		25	50	uA
I_{OP}	Typical operation current	$F_{OP}=40kHz$		1.5		mA
FB section						
V_{FB}	FB voltage threshold			1		V
$V_{FB-CLAMP}$	FB clamp voltage	$I_{FB}=2uA$		1.5		V
$T_{DEMAG-MIN}$	Minimum demagnetization time			4.2		uS
$\Delta V_{CS} / \Delta I_{FBUP}$	Line compensation ratio	(note 7)		1.1		mV/uA
Maximum Duty Cycle						
D_{MAX}	Maximum Duty Cycle			42		%
Over temperature protection						
T_{SD}	Thermal shutdown threshold			160		$^\circ C$
T_{SD-HYS}	Thermal shutdown hysteresis			30		$^\circ C$
Gate Driver section						
I_{SOURCE}	Gate sourcing current			100		mA
R_{DSOFF}	Gate switch off resistor			30		Ω

Note5 : production testing of the chip is performed at $25^\circ C$.

Note6: the maximum and minimum parameters specified are guaranteed by test, the typical value are guaranteed by design, characterization and statistical analysis

Note7: refer to application information

Application Information

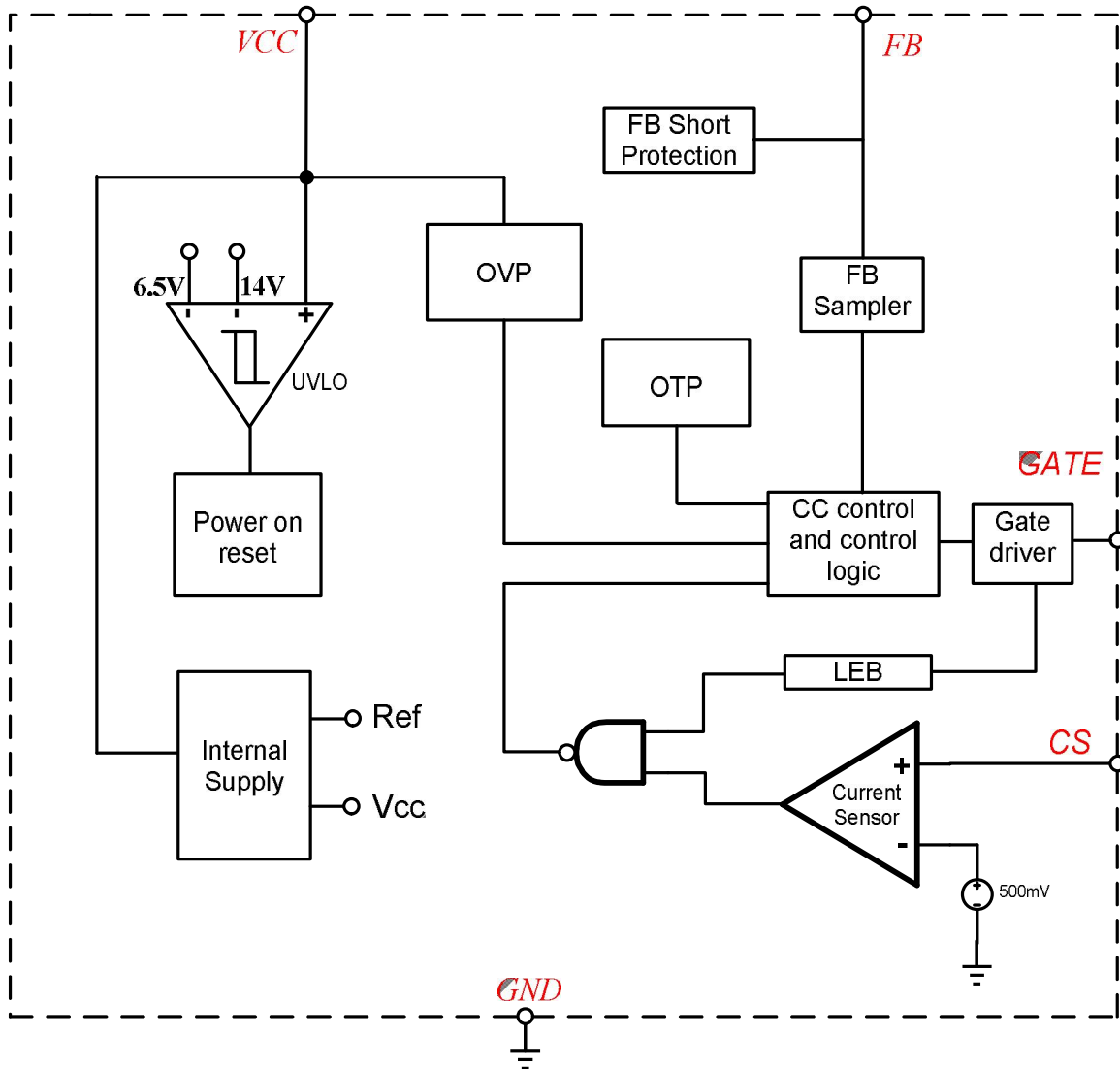


Figure 2 block diagram

BP3105 is a high performance AC/DC controller specially designed for LED lighting. The patented control method dramatically saves the system cost and size by eliminating all secondary feedback circuitry and loop control compensation circuitry. No TL431 and opto-coupler is required to get excellent constant current control.

1 Start up

The start-up current is designed to be low, only 25uA start-up current is required for the device. After the system is powered on, the Vcc capacitor is charged

through the start-up resistor, once the Vcc voltage reaches the start-up threshold (14V typically), the chip start to operate. During normal operation, The Vcc is supplied via the auxiliary winding.

2 Constant current control

Cycle-by-Cycle current limiting is offered .The CS is connected to non-inverting inputs of the current sense comparator. The inverting terminal of comparator is tied to an internal 500mV reference. The comparator is constantly comparing the CS voltage to the internal 500mV. Once the output of

comparator goes high, the GATE output goes low.

The primary peak current is given by:

$$I_P = 500(\text{mV}) / R_{CS} (\text{mA})$$

The outputs of the comparators also include a typical 500ns (typ.) blanking time which prevents spurious turn-offs of the external MOSFET due to the turn-on spike normally.

The LED current is decided by the internal constant current control, which is given by:

$$I_{OUT} = \frac{I_P}{4} \times \frac{N_P}{N_S}$$

N_P is turns of transformer primary winding, N_S is turns of transformer secondary winding, I_P is the peak current of the primary side

3 Feedback network

The chip senses the output voltage and output current through the feedback network. The FB sense threshold is 1V. We recommend to set the feedback upper resistor to 300K Ω ~750K Ω to improve system efficiency.

The FB upper resistor is also used to adjust the V_{cs} threshold to compensation the input voltage change. It is given by:

$$\Delta V_{CS} = -1.1 \times 10^6 * \frac{V_{AUX}}{R_{FBH}} (\text{mV})$$

R_{FBH} is the upper resistor of the feedback network

4 Gate drive

The driver structure of BP3105 is optimized. Too strong drive current results in bad EMI while too weak driver cause big switch loss of power MOSFET. A good trade-off is achieved in the internal driver with suitable strength.

5 Operation switching frequency

The system work in DCM mode, the maximum duty cycle is 42%; Normally the system frequency is recommended to be set between 40KHz-48KHz to

get better EMI result. The maximum operating frequency should be lower than 80KHz, while the minimum system frequency should be higher than 20KHz. The operation frequency is given by:

$$f = \frac{N_P^2 * V_{LED}}{8 * N_S^2 * L_P * I_{LED}}$$

L_P is inductance of primary winding.

The chip limits the maximum system frequency and the minimum system frequency to ensure the system stability.

6 Protection control

Good power supply system reliability is achieved with its rich protection features.

Once the V_{cc} voltage is higher than 16V (typ.), the device shut down the MOSFET and goes into auto-start mode until the over-voltage is removed.

Another 19V clamp circuitry is offered to prevent the chip from damage on abnormal situation.

The thermal shutdown circuitry senses the die temperature. The threshold is set at 160 $^{\circ}\text{C}$ typical with a 30 $^{\circ}\text{C}$ hysteresis. When the die temperature rises above this threshold (160 $^{\circ}\text{C}$) the power MOSFET is disabled and remains disabled until the die temperature falls by 30 $^{\circ}\text{C}$, at which point the MOSFET is re-enabled.

Once LED short circuit or LED open circuit is detected, the system enters into low standby mode in which the system consumes very low power and the output status is monitored continuously. The chip goes back to normal operation automatically after the abnormal situation is removed.

7 PCB Layout

The following rules should be followed in BP3105 PCB layout.

Bypass capacitor

The V_{cc} bypass capacitor should be as close to the V_{cc} pin as possible.

Current sense (CS) resistor

The CS resistor ground should be as close to the chip ground as possible to reduce the ground noise coupled, which improves the quality of the sensing signal.

Current path area

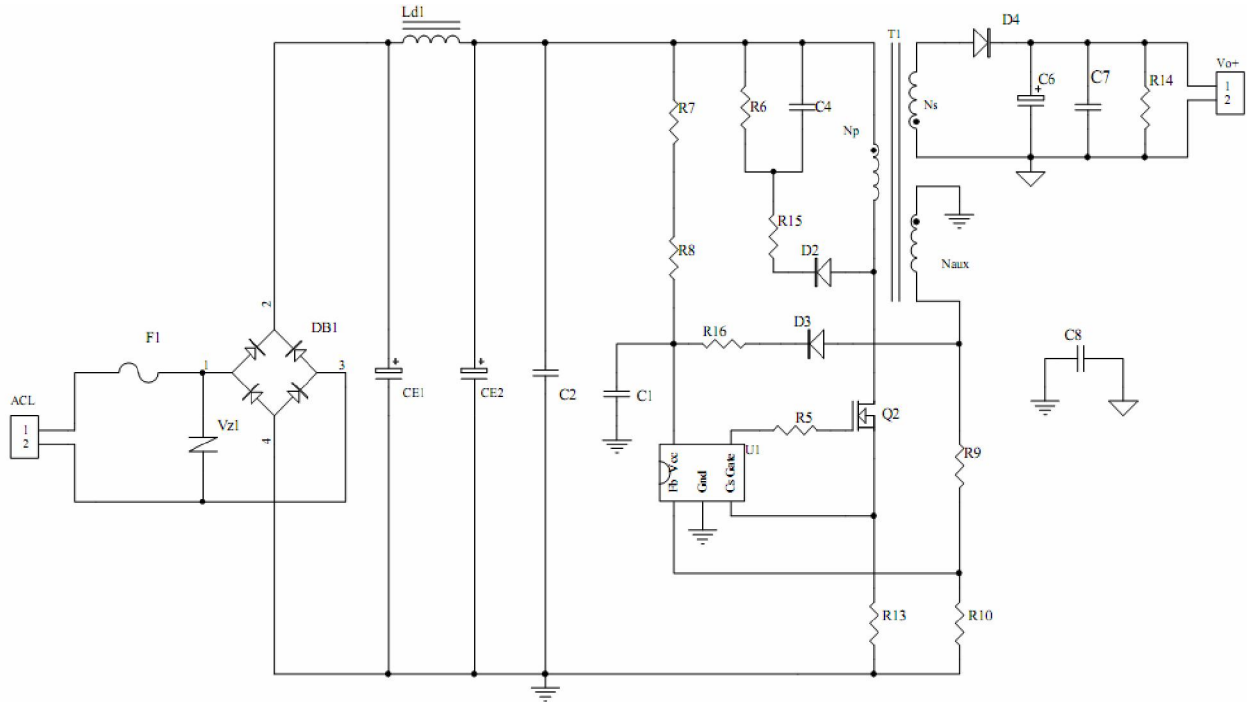
The area of current path should be as small as possible to reduce EMI radiation, such as the primary current loop from the primary inductor to the MOSFET and the secondary current loop from the secondary inductor to the output rectifier.

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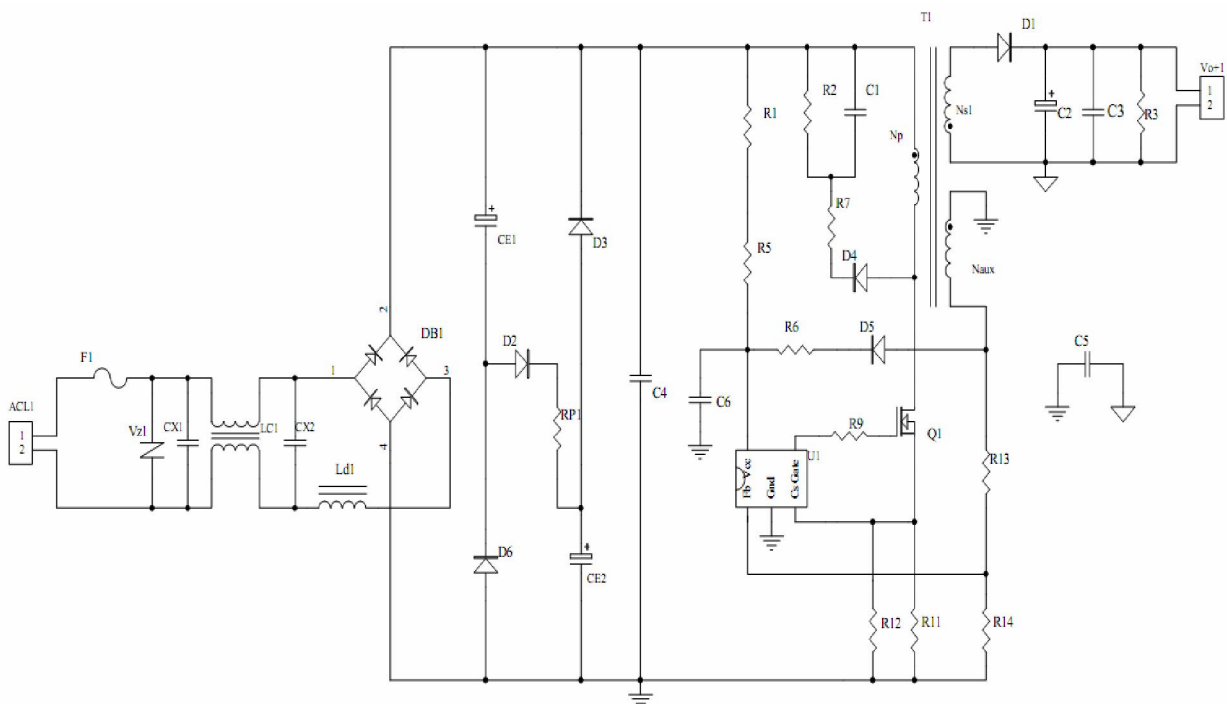
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Application Example

Input voltage range: 85V-264V



Input voltage range: 85V-264V with passive power factor correction (PPFC)



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Package Information

