National Semiconductor

LP2950/A-XX and LP2951/A-XX Series of Adjustable Micropower Voltage Regulators

General Description

The LP2950 and LP2951 are micropower voltage regulators with very low quiescent current (75 μA typ.) and very low dropout voltage (typ. 40 mV at light loads and 380 mV at 100 mA). They are ideally suited for use in battery-powered systems. Furthermore, the quiescent current of the LP2950/LP2951 increases only slightly in dropout, prolonging battery life.

The LP2950-5.0 in the popular 3-pin TO-92 package is pincompatible with older 5V regulators. The 8-lead LP2951 is available in plastic, ceramic dual-in-line, or metal can packages and offers additional system functions.

One such feature is an error flag output which warns of a low output voltage, often due to falling batteries on the input. It may be used for a power-on reset. A second feature is the logic-compatible shutdown input which enables the regulator to be switched on and off. Also, the part may be pin-strapped for a 5V, 3V, or 3.3V output (depending on the version), or programmed from 1.24V to 29V with an external pair of resistors.

Careful design of the LP2950/LP2951 has minimized all contributions to the error budget. This includes a tight initial

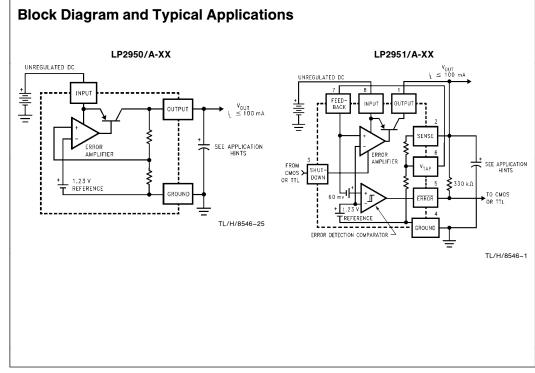
tolerance (.5% typ.), extremely good load and line regulation (.05% typ.) and a very low output voltage temperature coefficient, making the part useful as a low-power voltage reference.

Features

- 5V, 3V, and 3.3V versions available
- High accuracy output voltage
- Guaranteed 100 mA output current
- Extremely low quiescent current
- Low dropout voltage
- Extremely tight load and line regulation
- Very low temperature coefficient
- Use as Regulator or Reference
- Needs minimum capacitance for stability
- Current and Thermal Limiting

LP2951 versions only

- Error flag warns of output dropout
- Logic-controlled electronic shutdown
- Output programmable from 1.24 to 29V

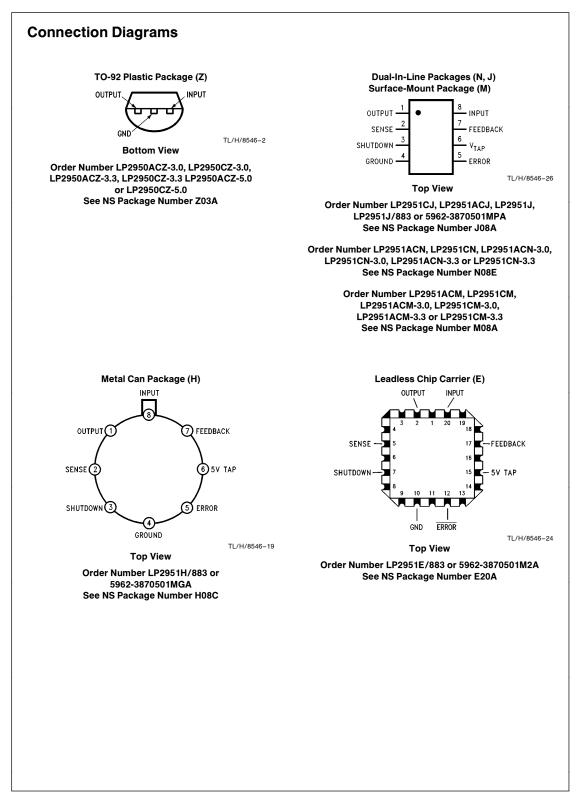


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LP2950/A-XX and LP2951/A-XX

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Ordering Information

Package		Temperature		
Fackage	3.0V	3.3V	5.0V	(°C)
TO-92 (Z)	LP2950ACZ-3.0 LP2950CA-3.0	LP2950ACZ-3.3 LP2950CZ-3.3	LP2950ACZ-5.0 LP2950CZ-5.0	-40 < T _J < 125
N (N-08E)	LP2951ACN-3.0 LP2951CN-3.0	LP2951ACN-3.3 LP2951CN-3.3	LP2951ACN LP2950CN	$-40 < T_{J} < 125$
M (M08A)	LP2951ACM-3.0 LP2951CM-3.0	LP2951ACM-3.3 LP2951CM-3.3	LP2951ACM LP2951CM	$-40 < T_{J} < 125$
J (J08A)			LP2951ACJ LP2951CJ	$-40 < T_{J} < 125$
			LP2951J LP2951J/883 5926-3870501MPA	—55 < Т _Ј < 150
H (H08C)			LP2951H/883 5962-3870501MGA	$-55 < T_{J} < 150$
E (E20A)			LP2951E/883 5962-3870501M2A	$-55 < T_{J} < 150$

Absolute Maximum Rat	ings		
If Military/Aerospace specified de	vices are required,	Input Supply Voltage	-0.3 to $+30V$
please contact the National Ser Office/Distributors for availability a		Feedback Input Voltage (Notes 9 and 10)	-1.5 to $+30V$
Power Dissipation	Internally Limited	Shutdown Input Voltage	-0.3 to +30 V
Lead Temp. (Soldering, 5 seconds)	260°C	(Note 9)	
Storage Temperature Range	-65° to +150°C	Error Comparator Output	
Operating Junction Temperature Rang	e (Note 8)	Voltage (Note 9)	-0.3 to $+30V$
LP2951 LP2950AC-XX, LP2950C-XX,	-55° to +150°C	ESD Rating is to be determined.	

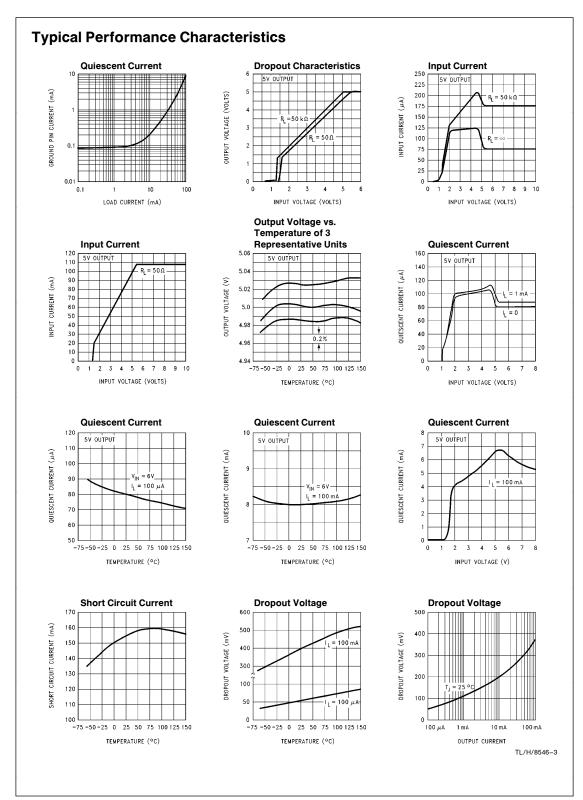
LP2951AC-XX, LP2951C-XX -40° to +125°C

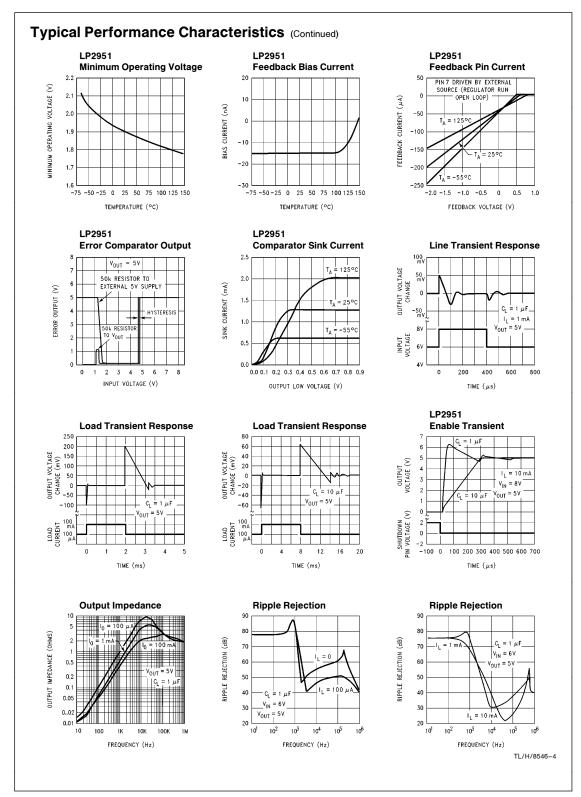
Electrical Characteristics (Note 1)

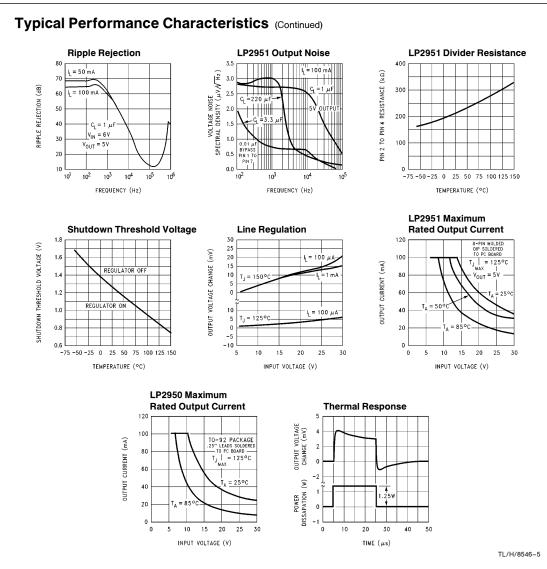
	Conditions (Note 2)	LP2951			_P2950A(_P2951A(LP2950C-XX LP2951C-XX			
Parameter		Тур	Tested Limit (Notes 3, 16)	Тур	Tested Limit (Note 3)	Design Limit (Note 4)	Тур	Tested Limit (Note 3)	Design Limit (Note 4)	Units
3V VERSIONS (Note 17)	-								-	
Output Voltage	$T_{J} = 25^{\circ}C$	3.0	3.015 2.985	3.0	3.015 2.985		3.0	3.030 2.970		V max V min
	$-25^{\circ}C \leq T_J \leq 85^{\circ}C$	3.0		3.0		3.030 2.970	3.0		3.045 2.955	V max V min
	Full Operating Temperature Range	3.0	3.036 2.964	3.0		3.036 2.964	3.0		3.060 2.940	V max V min
Output Voltage	$\begin{array}{l} 100 \; \mu A \leq I_L \leq 100 \; mA \\ T_J \leq T_{JMAX} \end{array}$	3.0	3.045 2.955	3.0		3.042 2.958	3.0		3.072 2.928	V max V min
3.3V VERSIONS (Note 1	7)									
Output Voltage	$T_{J} = 25^{\circ}C$	3.3	3.317 3.284	3.3	3.317 3.284		3.3	3.333 3.267		V max V min
	$-25^{\circ}C \le T_{J} \le 85^{\circ}C$	3.3		3.3		3.333 3.267	3.3		3.350 3.251	V max V min
	Full Operating Temperature Range	3.3	3.340 3.260	3.3		3.340 3.260	3.3		3.366 3.234	V max V min
Output Voltage	$\begin{array}{l} 100 \ \mu A \leq I_L \leq 100 \ \text{mA} \\ T_J \leq T_{JMAX} \end{array}$	3.3	3.350 3.251	3.3		3.346 3.254	3.3		3.379 3.221	V max V min
5V VERSIONS (Note 17)			•		•	•		•		
Output Voltage	$T_{J} = 25^{\circ}C$	5.0	5.025 4.975	5.0	5.025 4.975		5.0	5.05 4.95		V max V min
	$-25^{\circ}C \le T_{J} \le 85^{\circ}C$	5.0		5.0		5.05 4.95	5.0		5.075 4.925	V max V min
	Full Operating Temperature Range	5.0	5.06 4.94	5.0		5.06 4.94	5.0		5.1 4.9	V max V min
Output Voltage	100 μ A \leq I _L \leq 100 mA T _J \leq T _{JMAX}	5.0	5.075 4.925	5.0		5.075 4.925	5.0		5.12 4.88	V max V min
ALL VOLTAGE OPTION	S									
Output Voltage Temperature Coefficient	(Note 12)	20	120	20		100	50		150	ppm/°C
Line Regulation (Note 14)	$\begin{array}{l} (V_ONOM + \ 1)V \leq V_{in} \leq 30V \\ (Note \ 15) \end{array}$	0.03	0.1 0.5	0.03	0.1	0.2	0.04	0.2	0.4	% max % max
Load Regulation (Note 14)	$100 \ \mu A \leq I_L \leq 100 \ mA$	0.04	0.1 0.3	0.04	0.1	0.2	0.1	0.2	0.3	% max % max

Parameter	Conditions (Note 2)	LP2951		LP2950AC-XX LP2951AC-XX			LP2950C-XX LP2951C-XX			
		Тур	Tested Limit (Notes 3, 16)	Тур	Tested Limit (Note 3)	Design Limit (Note 4)	Тур	Tested Limit (Note 3)	Design Limit (Note 4)	Units
ALL VOLTAGE OPTION	S (Continued)									
Dropout Voltage (Note 5)	I _L = 100 μA	50	80 150	50	80	150	50	80	150	mV max mV max
	$I_L = 100 \text{ mA}$	380	450 600	380	450	600	380	450	600	mV max mV max
Ground Current	I _L = 100 μA	75	120 140	75	120	140	75	120	140	μA max μA max
	I _L = 100 mA	8	12 14	8	12	14	8	12	14	mA max mA max
Dropout Ground Current	$\label{eq:Vin} \begin{split} V_{in} &= (V_O NOM - 0.5) V \\ I_L &= 100 \; \mu A \end{split}$	110	170 200	110	170	200	110	170	200	μA max μA max
Current Limit	$V_{out} = 0$	160	200 220	160	200	220	160	200	220	mA max mA max
Thermal Regulation	(Note 13)	0.05	0.2	0.05	0.2		0.05	0.2		%/W max
Output Noise, 10 Hz to 100 KHz	$C_L = 1 \ \mu F$ (5V Only)	430		430			430			μV rms
	$C_L = 200 \ \mu F$	160		160			160			μV rms
	$C_L = 3.3 \mu\text{F}$ (Bypass = 0.01 μF Pins 7 to 1 (LP2951))	100		100			100			μV rms
8-PIN VERSIONS ONLY		LP2951		LP2951AC-XX			LP2951C-XX			
Reference Voltage		1.235	1.25 1.26 1.22	1.235	1.25 1.22	1.26	1.235	1.26 1.21	1.27	V max V max V min
			1.2			1.2			1.2	V min
Reference Voltage	(Note 7)		1.27 1.19			1.27 1.19			1.285 1.185	V max V min
Feedback Pin Bias Current		20	40 60	20	40	60	20	40	60	nA max nA max
Reference Voltage Temperature Coefficient	(Note 12)	20		20			50			ppm/°C
Feedback Pin Bias Current Temperature Coefficient		0.1		0.1			0.1			nA/°C
Error Comparator										
Output Leakage Current	$V_{OH} = 30V$	0.01	1 2	0.01	1	2	0.01	1	2	μA max μA max
Output Low Voltage	$V_{in} = (V_O NOM - 0.5)V$ $I_{OL} = 400 \ \mu A$	150	250 400	150	250	400	150	250	400	mV max mV max
Upper Threshold Voltage	(Note 6)	60	40 25	60	40	25	60	40	25	mV min mV min
Lower Threshold Voltage	(Note 6)	75	95 140	75	95	140	75	95	140	mV max mV max
Hysteresis	(Note 6)	15		15			15			mV

Parameter			LP2951		LP2951AC	-XX		-XX		
	Conditions (Note 2)	Тур	Tested Limit (Notes 3, 16)	Тур	Tested Limit (Note 3)	Design Limit (Note 4)	Тур	Tested Limit (Note 3)	Design Limit (Note 4)	Uni
PIN VERSIONS ON	LY (Continued)									
nutdown Input										
nput ogic ′oltage	Low (Regulator ON) High (Regulator OFF)	1.3	0.6 2.0	1.3		0.7 2.0	1.3		0.7 2.0	V Vm: Vm
Shutdown Pin nput Current	V _{shutdown} = 2.4V	30	50 100	30	50	100	30	50	100	μAn μAn
	V _{shutdown} = 30V	450	600 750	450	600	750	450	600	750	μAn μAn
Regulator Output Current in Shutdown	(Note 11)	3	10 20	3	10	20	3	10	20	μAn μAn
Note 8: The junction-to- resistance of the 8-pin I board. Thermal resistanc (M) package is 160°C/W Note 9: May exceed inp Note 10: When used in ground. Note 11: V _{shutdown} \geq 2 Note 12: Output or refe Note 13: Thermal regula effects. Specifications a Note 14: Regulation is in covered under the spec Note 15: Line regulation Typical Performance Ch Note 16: A Military RETS or J may also be procur Note 17: All LP2950 dev	$V_{in} - 1V$), 2.3V $\leq V_{in} \leq 30V$, ambient thermal resistance of DIP packages is 105°C/W for 1 ce for the metal can (H) is 160° λ . Thermal resistance for the k ut supply voltage. dual-supply systems where the V, $V_{in} \leq 30V$, $V_{out} = 0$, Feed rence voltage temperature coe- tition is defined as the change i rea for a 50 mA load pulse at V neasured at constant junction n dification for thermal regulation. for the LP2951 is tested at 15 aracteristics for line regulation δ spec is available on request. <i>i</i> ded as Standard Military Drawir <i>i</i> ces have the nominal output v sist two digits, but the 5V version	the TO-S the mold C/W jur eadless e output back pin officient i in output 'IN = 30 cemperat 0°C for I versus At time c ng Spec voltage c	I2 package is 180° C ed plastic (N) and tction to ambient an chip carrier (E) pack terminal sees loads tied to V _{TAP} . s defined as the wo voltage at a time T V (1.25W pulse) for ure, using pulse tes L = 1 mA. For IL = temperature and loc f printing, the LP29! # 5982-8870501MC	C/W with 130° C/W $d 20^{\circ}$ C	0.4'' leads a / for the cerd W junction to 55° C/W juncti d to a negativ voltage charge in pow- ms. a low duty cy and T _J = 12 nt. spec complie or MPA. of the part num	ip (J) junction case. Junction on to ambien we supply, the use divided b ver dissipation ycle. Changes 25°C, line regu- ed with the bo- mber. In the L	n to amb on to ami t and 24 e output y the tota n is appl s in outpu ulation is uldface lin .P2951 p	ient when so bient thermal °C/W junction voltage shoul al temperatur ied, excluding ut voltage due guaranteed t mits in this col roducts, the 3	Idered directly resistance fo n to case. d be diode-cl e range. J load or line r e to heating e by design to 0 lumn. The LP2 3.0V and 3.3V	y to a F r the S. amped regulation ffects a .2%. S 2951H, v versio







Application Hints

EXTERNAL CAPACITORS

A 1.0 μF (or greater) capacitor is required between the output and ground for stability at output voltages of 5V or more. At lower output voltages, more capacitance is required (2.2 μF or more is recommended for 3V and 3.3V versions). Without this capacitor the part will oscillate. Most types of tantalum or aluminum electrolytics work fine here; even film types work but are not recommended for reasons of cost. Many aluminum electrolytics have electrolytes that freeze at about $-30^\circ C$, so solid tantalums are recommended for operation below $-25^\circ C$. The important parameters of the capacitor are an ESR of about 5 Ω or less and a resonant frequency above 500 kHz. The value of this capacitor may be increased without limit.

At lower values of output current, less output capacitance is required for stability. The capacitor can be reduced to

0.33 μ F for currents below 10 mA or 0.1 μ F for currents below 1 mA. Using the adjustable versions at voltages below 5V runs the error amplifier at lower gains so that *more* output capacitance is needed. For the worst-case situation of a 100 mA load at 1.23V output (Output shorted to Feedback) a 3.3 μ F (or greater) capacitor should be used.

Unlike many other regulators, the LP2950 will remain stable and in regulation with no load in addition to the internal voltage divider. This is especially important in CMOS RAM keep-alive applications. When setting the output voltage of the LP2951 versions with external resistors, a minimum load of 1 μ A is recommended.

A 1 μ F tantalum or aluminum electrolytic capacitor should be placed from the LP2950/LP2951 input to ground if there is more than 10 inches of wire between the input and the AC filter capacitor or if a battery is used as the input.

Application Hints (Continued)

Stray capacitance to the LP2951 Feedback terminal can cause instability. This may especially be a problem when using high value external resistors to set the output voltage. Adding a 100 pF capacitor between Output and Feedback and increasing the output capacitor to at least 3.3 μ F will fix this problem.

ERROR DETECTION COMPARATOR OUTPUT

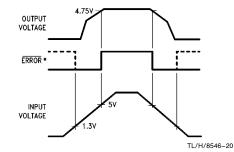
The comparator produces a logic low output whenever the LP2951 output falls out of regulation by more than approximately 5%. This figure is the comparator's built-in offset of about 60 mV divided by the 1.235 reference voltage. (Refer to the block diagram in the front of the datasheet.) This trip level remains "5% below normal" regardless of the programmed output voltage of the 2951. For example, the error flag trip level is typically 4.75V for a 5V output or 11.4V for a 12V output. The out of regulation condition may be due either to low input voltage, current limiting, or thermal limiting. Figure 1 below gives a timing diagram depicting the ERROR signal and the regulated output voltage as the LP2951 input is ramped up and down. For 5V versions, the ERROR signal becomes valid (low) at about 1.3V input. It goes high at about 5V input (the input voltage at which $V_{OUT} = 4.75$). Since the LP2951's dropout voltage is load-dependent (see

curve in typical performance characteristics), the **input** voltage trip point (about 5V) will vary with the load current. The **output** voltage trip point (approx. 4.75V) does not vary with load.

The error comparator has an open-collector output which requires an external pullup resistor. This resistor may be returned to the output or some other supply voltage depending on system requirements. In determining a value for this resistor, note that while the output is rated to sink 400 μ A, this sink current adds to battery drain in a low battery condition. Suggested values range from 100k to 1 M Ω . The resistor is not required if this output is unused.

PROGRAMMING THE OUTPUT VOLTAGE (LP2951)

The LP2951 may be pin-strapped for the nominal fixed output voltage using its internal voltage divider by tying the output and sense pins together, and also tying the feedback and V_{TAP} pins together. Alternatively, it may be programmed for any output voltage between its 1.235V reference and its 30V maximum rating. As seen in *Figure 2*, an external pair of resistors is required.



*When $V_{IN} \leq 1.3V$, the error flag pin becomes a high impedance, and the error flag voltage rises to its pull-up voltage. Using V_{OUT} as the pull-up voltage (see Figure 2), rather than an external 5V source, will keep the error flag voltage under 1.2V (typ.) in this condition. The user may wish to divide down the error flag voltage using equal-value resistors (10 k Ω suggested), to ensure a low-level logic signal during any fault condition, while still allowing a valid high logic level during normal operation.

FIGURE 1. ERROR Output Timing

The complete equation for the output voltage is

$$V_{OUT} = V_{REF} \bullet \left(1 + \frac{R_1}{R_2}\right) + I_{FB}R_1$$

where V_{REF} is the nominal 1.235 reference voltage and I_{FB} is the feedback pin bias current, nominally -20 nA. The minimum recommended load current of 1 μ A forces an upper limit of 1.2 $M\Omega$ on the value of R₂, if the regulator must work with no load (a condition often found in CMOS in standby). I_{FB} will produce a 2% typical error in V_{OUT} which may be eliminated at room temperature by trimming R₁. For better accuracy, choosing R₂ = 100k reduces this error to 0.17% while increasing the resistor program current to 12 μ A. Since the LP2951 typically draws 60 μ A at no load with Pin 2 open-circuited, this is a small price to pay.

REDUCING OUTPUT NOISE

In reference applications it may be advantageous to reduce the AC noise present at the output. One method is to reduce the regulator bandwidth by increasing the size of the output capacitor. This is the only way noise can be reduced on the 3 lead LP2950 but is relatively inefficient, as increasing the capacitor from 1 μF to 220 μF only decreases the noise from 430 μV to 160 μV rms for a 100 kHz bandwidth at 5V output.

Noise can be reduced fourfold by a bypass capacitor accross $\mathsf{R}_1,$ since it reduces the high frequency gain from 4 to unity. Pick

$$C_{BYPASS} \cong \frac{1}{2\pi B_1 \bullet 200 \text{ Hz}}$$

1

or about 0.01 $\mu F.$ When doing this, the output capacitor must be increased to 3.3 μF to maintain stability. These changes reduce the output noise from 430 μV to 100 μV rms for a 100 kHz bandwidth at 5V output. With the bypass capacitor added, noise no longer scales with output voltage so that improvements are more dramatic at higher output voltages.

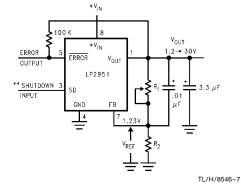
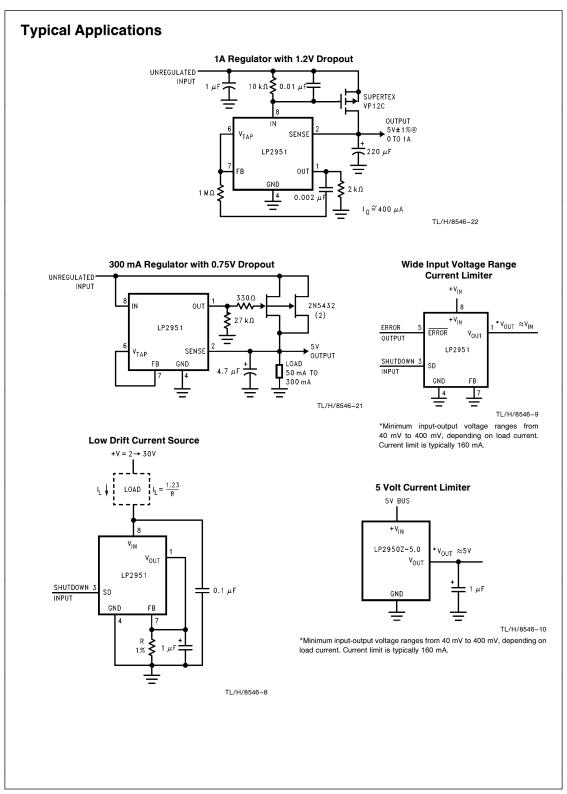


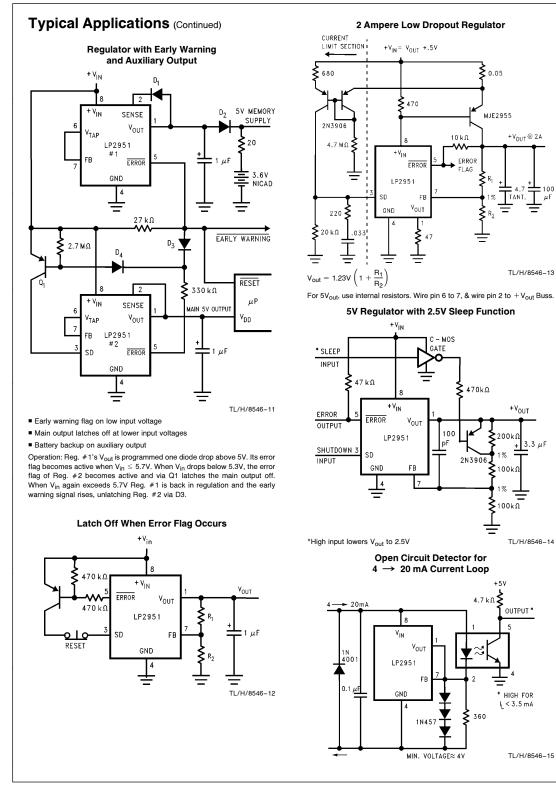
FIGURE 2. Adjustable Regulator

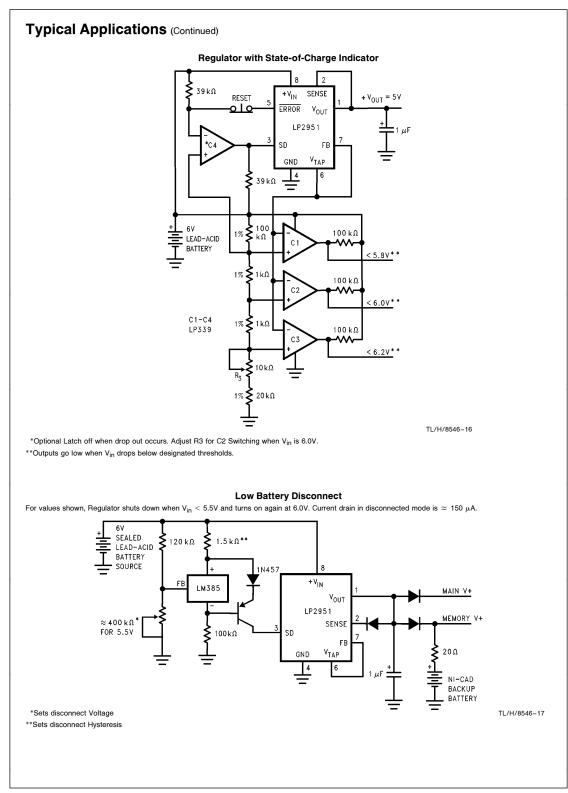
$$V_{out} = V_{Ref} \left(1 + \frac{R_1}{R_2} \right)$$

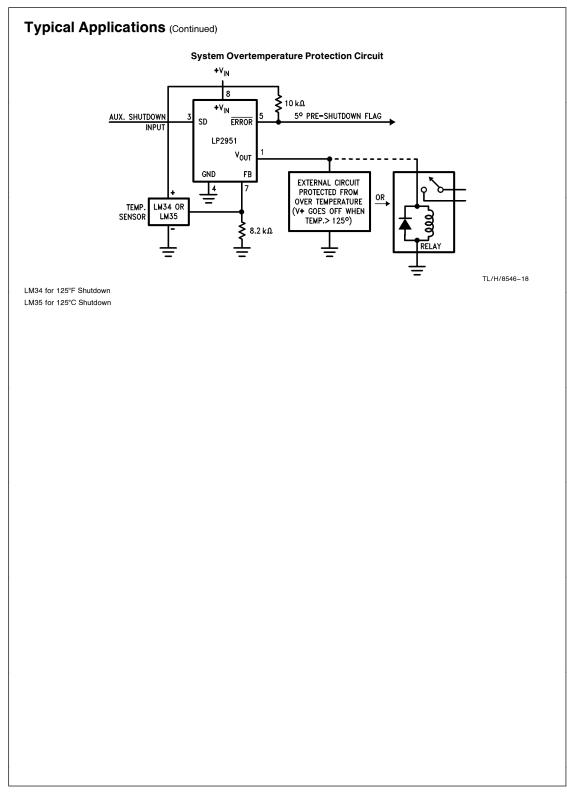
 $^{**}\mbox{Drive with TTL-high to shut down. Ground or leave open if shutdown feature is not to be used.$

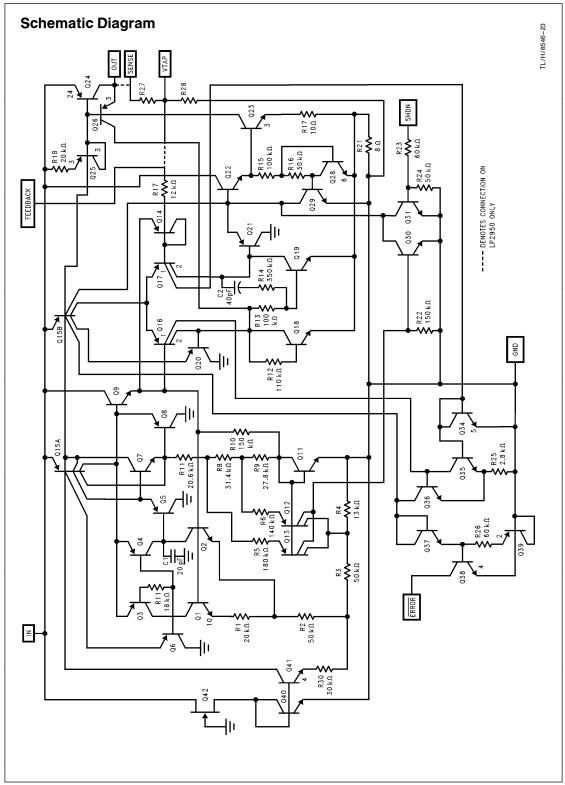
Note: Pins 2 and 6 are left open.

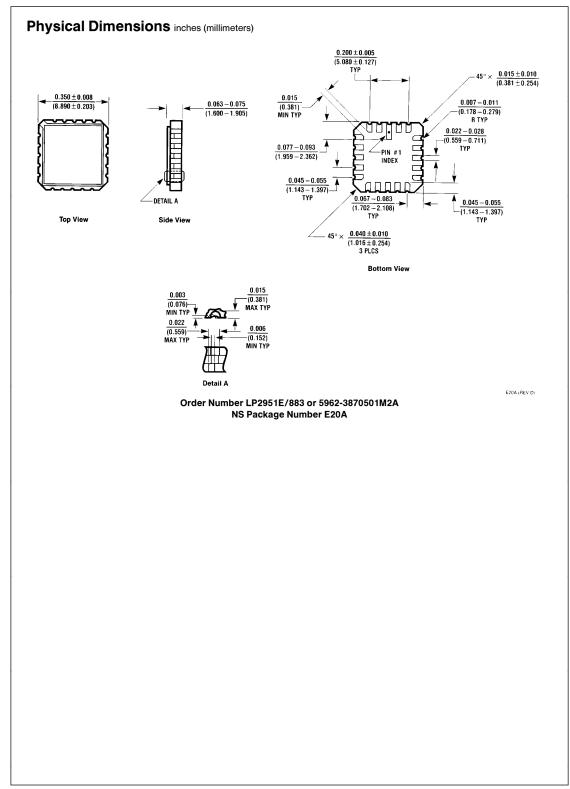


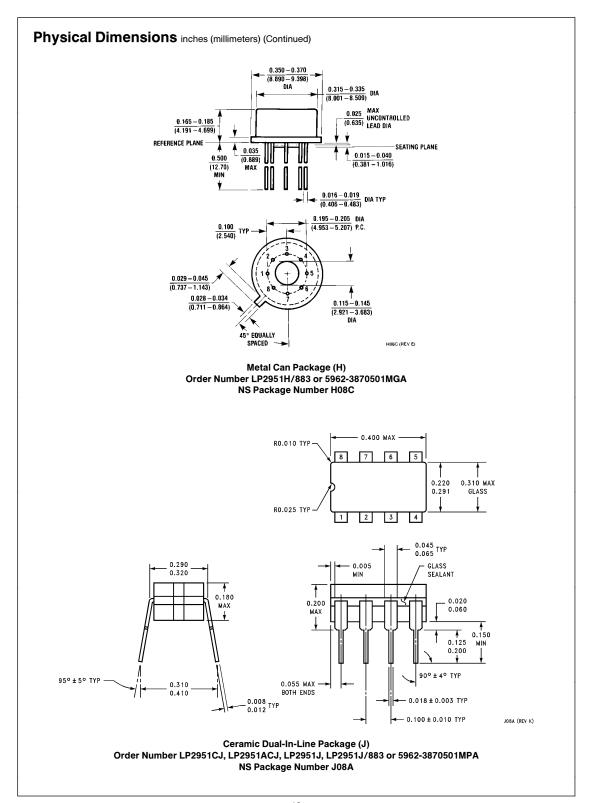


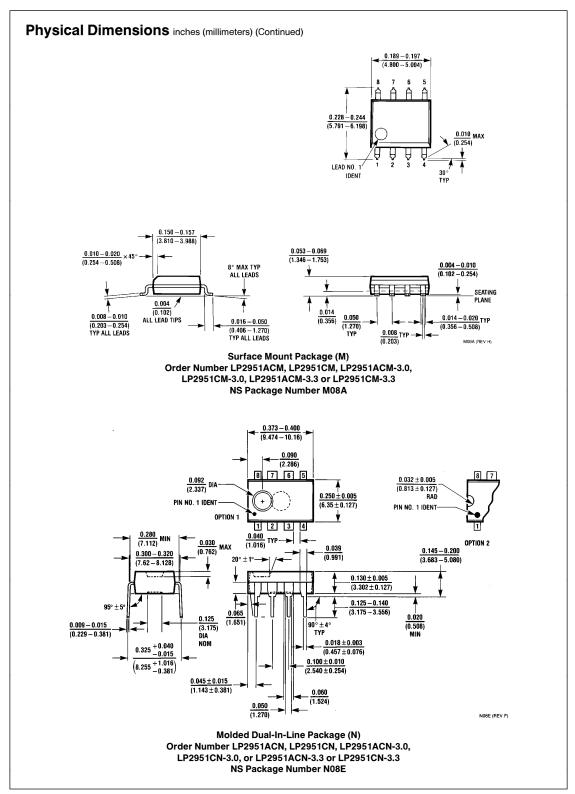




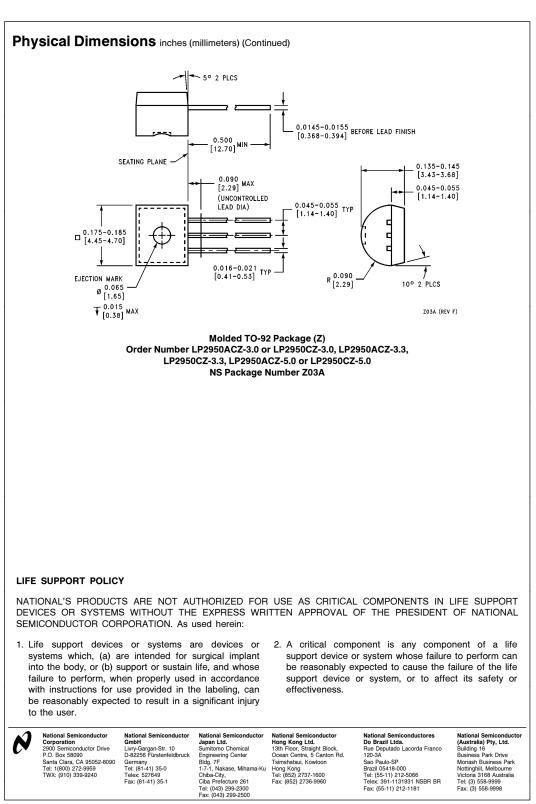












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